

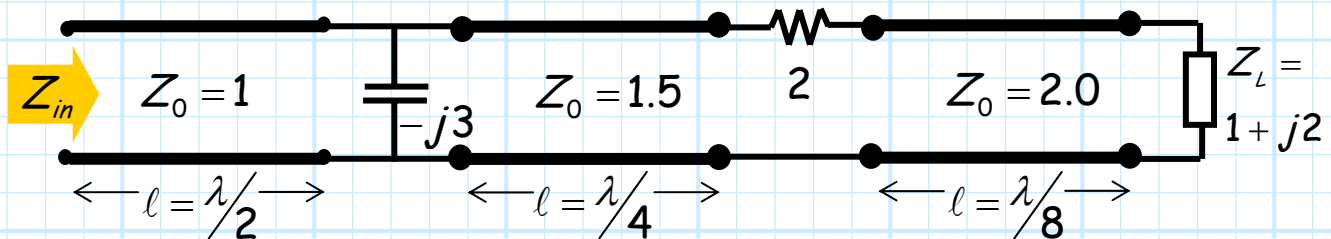
Note also for this case ( the electrically short transmission line), the voltage and current at each end of the transmission line are approximately the **same!**

$$V(z = -l) \approx V(z = 0) \quad \text{and} \quad I(z = -l) \approx I(z = 0) \quad \text{if} \quad l \ll \lambda$$

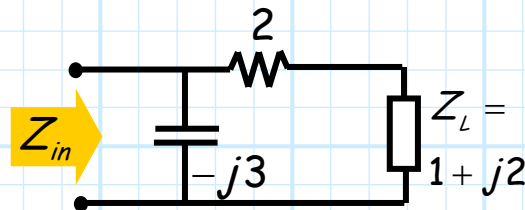
If  $l \ll \lambda$ , our "wire" behaves **exactly** as it did in EECS 211 !

# Example: Input Impedance

Consider the following circuit:



If we **ignored** our new  $\mu$ -wave knowledge, we might **erroneously** conclude that the input impedance of this circuit is:

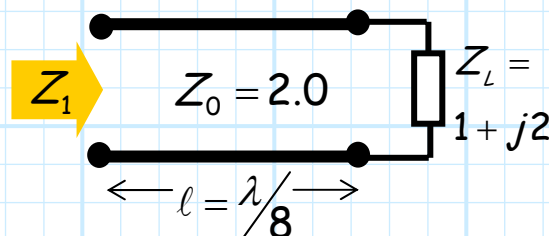


Therefore:

$$Z_{in} = \frac{-j3(2+1+j2)}{-j3+2+1+j2} = \frac{6-j9}{3-j} = 2.7 - j2.1$$

Of course, this is **not** the correct answer!

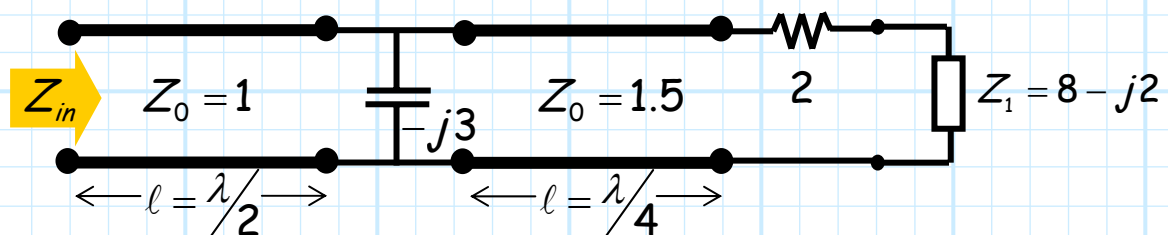
We must use our **transmission line theory** to determine an accurate value. Define  $Z_1$  as the input impedance of the last section:



we find that  $Z_1$  is :

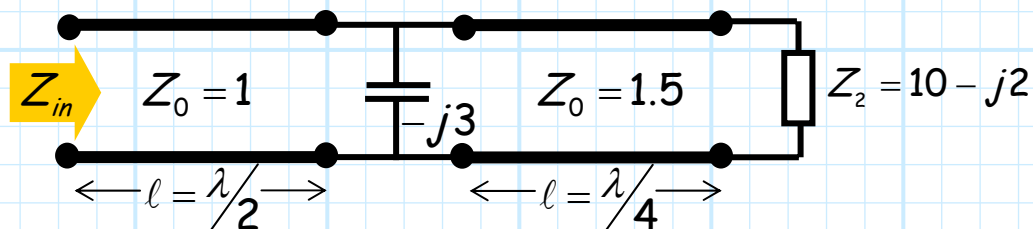
$$\begin{aligned} Z_1 &= Z_0 \left( \frac{Z_L \cos \beta l + j Z_0 \sin \beta l}{Z_0 \cos \beta l + j Z_L \sin \beta l} \right) \\ &= 2 \left( \frac{(1 + j2) \cos(\pi/4) + j 2 \sin(\pi/4)}{2 \cos(\pi/4) + j(1 + j2) \sin(\pi/4)} \right) \\ &= 2 \left( \frac{1 + j4}{j} \right) \\ &= 8 - j2 \end{aligned}$$

Therefore, our circuit now becomes:

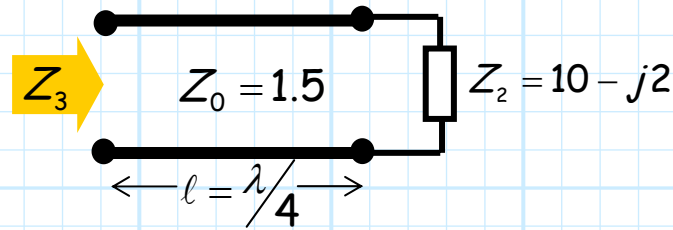


Note the resistor is in **series** with impedance  $Z_1$ . We can **combine** these two into one impedance defined as  $Z_2$ :

$$Z_2 = 2 + Z_1 = 2 + (8 - j2) = 10 - j2$$



Now let's define the input impedance of the **middle** transmission line section as  $Z_3$ :

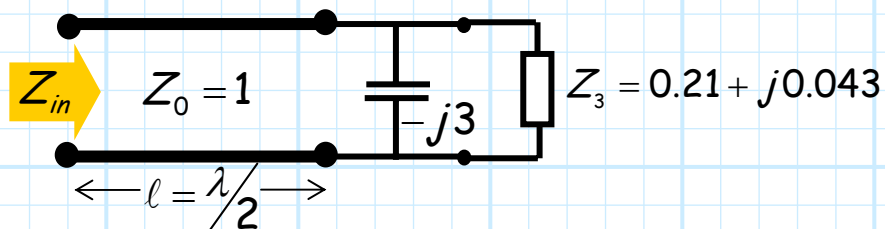


Note that this transmission line is a **quarter wavelength** ( $l = \lambda/4$ ). This is one of the **special** cases we considered earlier!

The input impedance  $Z_3$  is:

$$\begin{aligned} Z_3 &= \frac{Z_0^2}{Z_L} \\ &= \frac{Z_0^2}{Z_2} \\ &= \frac{1.5^2}{10 - j2} \\ &= 0.21 + j0.043 \end{aligned}$$

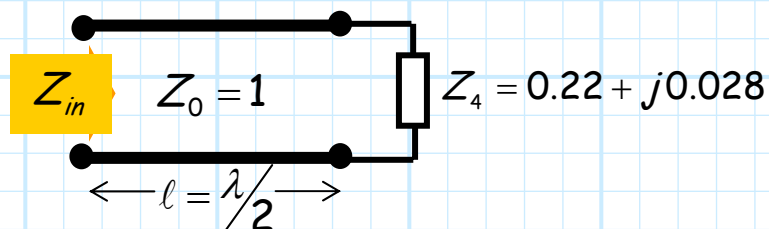
Thus, we can further **simplify** the original circuit as:



Now we find that the impedance  $Z_3$  is **parallel** to the capacitor. We can **combine** the two impedances and define the result as impedance  $Z_4$ :

$$\begin{aligned}
 Z_4 &= -j3 \parallel (0.21 + j0.043) \\
 &= \frac{-j3(0.21 + j0.043)}{-j3 + 0.21 + j0.043} \\
 &= 0.22 + j0.028
 \end{aligned}$$

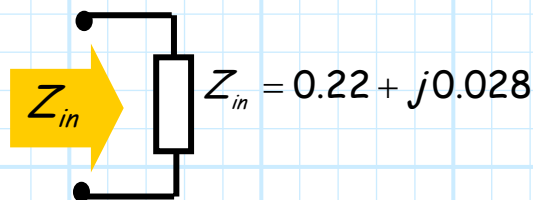
Now we are left with **this** equivalent circuit:



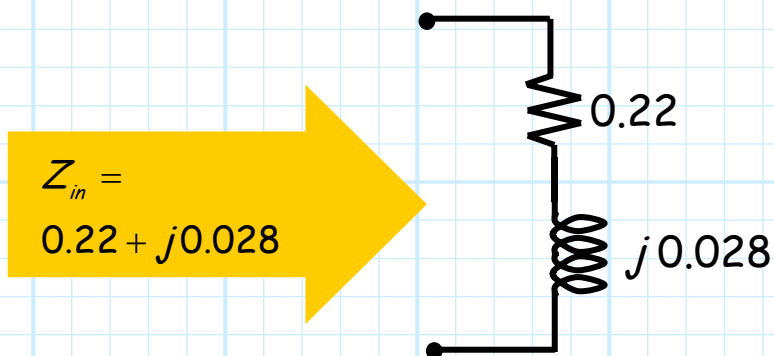
Note that the remaining transmission line section is a **half wavelength!** This is one of the **special** situations we discussed in a previous handout. Recall that the **input** impedance in this case is simply equal to the **load** impedance:

$$Z_{in} = Z_L = Z_4 = 0.22 + j0.028$$

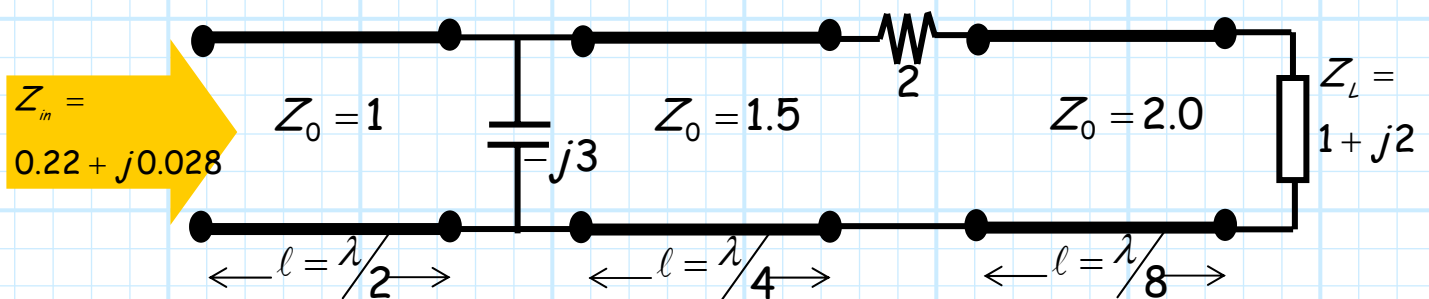
Whew! We are **finally** done. The **input impedance** of the original circuit is:



Note this means that **this** circuit:



and **this** circuit:



are precisely the **same**! They have **exactly** the same impedance, and thus they "behave" precisely the **same** way in any circuit (but **only** at frequency  $\omega_0$ !).